

# < DIPIPM > PSS75NE1CT

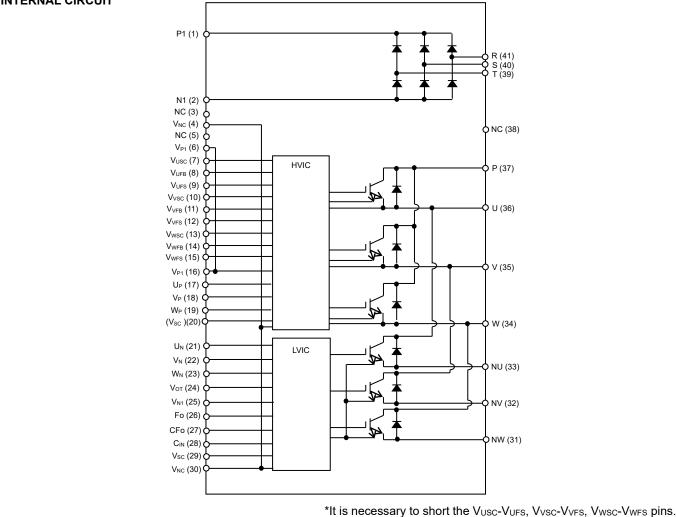
TRANSFER MOLDING TYPE INSULATED TYPE

# OUTLINEMAIN FUNCTIONImage: Strain of the st

### INTEGRATED DRIVE, PROTECTION AND SYSTEM CONTROL FUNCTIONS

- For P-side : Drive circuit, High voltage high-speed level shifting, Control supply under-voltage (UV) protection
- For N-side : Drive circuit, Control supply under-voltage protection (UV), Short circuit protection (SC)
- Fault signaling : Corresponding to SC fault (N-side IGBT), UV fault (N-side supply)
- Temperature output : Outputting LVIC temperature by analog signal
- Input interface : 3.3, 5V line (High Active)
- UL Recognized : UL1557 File E323585

#### **INTERNAL CIRCUIT**



# MITSUBISHI ELECTRIC CORPORATION

#### MAXIMUM RATINGS (T<sub>j</sub> = 25°C, unless otherwise noted)

#### **INVERTER PART**

Symbol	Parameter	Condition		Ratings	Unit
V <sub>cc</sub>	Supply voltage	Applied between P-NU, NV, NW		800	V
V <sub>CC(surge)</sub>	Supply voltage (surge)	Applied between P-NU, NV, NW		1000	V
V <sub>CES</sub>	Collector-emitter voltage			1200	V
±l <sub>c</sub>	Each IGBT collector current	T <sub>c</sub> = 25°C	(Note 1)	75	А
±I <sub>CP</sub>	Each IGBT collector current (peak)	$T_{\rm C}$ = 25°C, up to 1ms		112.5	А
T <sub>jop</sub>	Junction temperature	Continuous operation	(Note 2)	-30~+150	°C
T <sub>jmax</sub>	Maximum Junction temperature	Instantaneous event(overload)		175	°C

Note1: Pulse width and period are limited due to junction temperature.

Note2: The maximum junction temperature rating is 175°C. But for safe operation, it is recommended to limit the average junction temperature up to 150°C.

#### **CONVERTER PART**

Symbol	Parameter	Condition	Ratings	Unit
V <sub>RRM</sub>	Repetitive peak reverse voltage	etitive peak reverse voltage		V
lo	DC output current	3φ rectifying circuit	75	Α
I <sub>FSM</sub>	Surge forward current	Half sine at 60Hz, Peak value, Non-repetitive	600	Α
l <sup>2</sup> t	I <sup>2</sup> t for fusing	Value for 1 cycle of surge current	1440	A <sup>2</sup> s
Tj	Junction temperature	(Note 3)	-30~+150	°C

Note3: The maximum junction temperature rating is 150°C. But for safe operation, it is recommended to limit the average junction temperature up to 125°C.

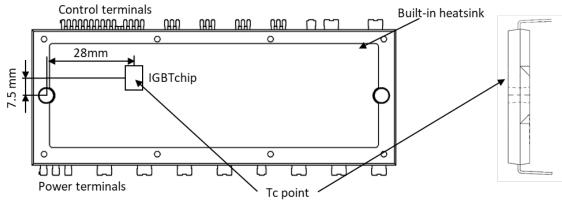
#### **CONTROL (PROTECTION) PART**

Symbol	Parameter	Condition	Ratings	Unit
VD	Control supply voltage	Applied between V <sub>P1</sub> -V <sub>PC</sub> , V <sub>N1</sub> -V <sub>NC</sub>	20	V
V <sub>DB</sub>	Control supply voltage	Applied between VUFB-VUFS, VVFB-VVFS, VWFB-VWFS	20	V
V <sub>IN</sub>	Input voltage	Applied between U <sub>P</sub> , V <sub>P</sub> , W <sub>P</sub> -V <sub>PC</sub> , U <sub>N</sub> , V <sub>N</sub> , W <sub>N</sub> -V <sub>NC</sub>	-0.5~V <sub>D</sub> +0.5	V
V <sub>FO</sub>	Fault output supply voltage	Applied between F <sub>O</sub> -V <sub>NC</sub>	-0.5~V <sub>D</sub> +0.5	V
I <sub>FO</sub>	Fault output current	Sink current at F <sub>o</sub> terminal	5	mA
V <sub>SC</sub>	Current sensing input voltage	Applied between CIN-V <sub>NC</sub>	-0.5~V <sub>D</sub> +0.5	V

#### TOTAL SYSTEM

Symbol	Parameter	Condition	Ratings	Unit
V <sub>CC(PROT)</sub>	Self protection supply voltage limit (Short circuit protection capability)	$V_D$ = 13.5~16.5V, Inverter Part T <sub>i</sub> = 150°C, non-repetitive, up to 2µs	800	V
Tc	Module case operation temperature	Tc measurement point is defined in Fig.1.	-30~+125	°C
T <sub>stg</sub>	Storage temperature		-40~+125	°C
V <sub>iso</sub>	Isolation voltage	60Hz, Sinusoidal, AC 1min, between connected all pins and heat sink plate	2500	V <sub>rms</sub>

#### Fig. 1: T<sub>C</sub> MEASUREMENT POINT



# THERMAL RESISTANCE

Symbol Parameter	Condition		Limits			
			Тур.	Max.	Unit	
R <sub>th(j-c)Q</sub>		Inverter IGBT part (per 1/6 module)		-	0.50	K/W
R <sub>th(j-c)F</sub>	Junction to case thermal resistance (Note 4)	Inverter FWDi part (per 1/6 module)	-	-	0.95	K/W
R <sub>th(j-c)R</sub>		Converter part (per 1/6module)			0.90	K/W

Note 4: Grease with good thermal conductivity and long-term endurance should be applied evenly with about +100µm~+200µm on the contacting surface of DIPIPM and heat-sink. The contacting thermal resistance between DIPIPM case and heat sink Rth(c-f) is determined by the thickness and the thermal conductivity of the applied grease. For reference, Rth(c-f) (per 1chip) is about 0.25K/W when the grease thickness is 20µm and the thermal conductivity is 1.0W/m·K

#### ELECTRICAL CHARACTERISTICS (T<sub>j</sub> = 25°C, unless otherwise noted) **INVERTER PART**

Baramatar	Condition			Limits			
Parameter		Condition		Тур.	Max.	Unit	
Collector-emitter saturation	V <sub>D</sub> =V <sub>DB</sub> =15V, V <sub>IN</sub> =5V,	T <sub>j</sub> = 25°C	-	1.50	1.90	V	
voltage	I <sub>C</sub> = 75A	T <sub>j</sub> = 125°C	-	1.80	2.30	v	
FWDi forward voltage	V <sub>IN</sub> = 0V, -I <sub>C</sub> = 75A	V <sub>IN</sub> = 0V, -I <sub>C</sub> = 75A		1.85	2.40	V	
			1.40	2.30	3.30	μs	
	$V_{cc} = 600V_{cc} = 15V_{cc}$	V	-	0.40	0.85	μs	
Switching times	I <sub>C</sub> = 75A, T <sub>j</sub> = 125°C, V <sub>IN</sub> =	V <sub>IN</sub> = 0⇔5V	-	2.70	3.80	μs	
	Inductive Load (upper-lo		-	0.30	0.95	μs	
			-	0.30	-	μs	
Collector-emitter cut-off current	V <sub>CE</sub> =V <sub>CES</sub>		-	-	1	mA	
	voltage FWDi forward voltage Switching times Collector-emitter	Collector-emitter saturation voltage $V_D = V_{DB} = 15V, V_{IN} = 5V,$ $I_C = 75A$ FWDi forward voltage $V_{IN} = 0V, -I_C = 75A$ Switching times $V_{CC} = 600V, V_D = V_{DB} = 15$ $I_C = 75A, T_j = 125°C, V_{IN} =$ Inductive Load (upper-loCollector-emitter $V_{CC} = V_{CD} = V_{CD}$	Collector-emitter saturation voltage $V_D = V_{DB} = 15V, V_{IN} = 5V, T_j = 25^{\circ}C$ $T_j = 125^{\circ}C$ FWDi forward voltage $V_{IN} = 0V, -I_C = 75A$ Switching times $V_{CC} = 600V, V_D = V_{DB} = 15V$ $I_C = 75A, T_j = 125^{\circ}C, V_{IN} = 0 \leftrightarrow 5V$ Inductive Load (upper-lower arm)Collector-emitter $V_{CC} = 5V_{CC} = 0$	Collector-emitter saturation voltage $V_D = V_{DB} = 15V, V_{IN} = 5V, \ I_C = 75A$ $T_J = 25^{\circ}C$ $T_J = 125^{\circ}C$ -FWDi forward voltage $V_{IN} = 0V, -I_C = 75A$ -FWDi forward voltage $V_{IN} = 0V, -I_C = 75A$ - $V_{CC} = 600V, V_D = V_{DB} = 15V$ $I_C = 75A, T_J = 125^{\circ}C, V_{IN} = 0 \leftrightarrow 5V$ Inductive Load (upper-lower arm)-Collector-emitter $V_{ac} = V_{ac}$ -	Parameter         Condition         Min.         Typ.           Collector-emitter saturation voltage $V_D = V_{DB} = 15V, V_{IN} = 5V, T_J = 25^{\circ}C$ -         1.50           FWDi forward voltage $V_{IN} = 0V, -I_C = 75A$ $T_J = 125^{\circ}C$ -         1.80           FWDi forward voltage $V_{IN} = 0V, -I_C = 75A$ $T_J = 125^{\circ}C$ -         1.85           Switching times $V_{CC} = 600V, V_D = V_{DB} = 15V$ $I.40$ 2.30 $V_{CC} = 600V, V_D = V_{DB} = 15V$ $I.40$ 2.30 $I_C = 75A, T_J = 125^{\circ}C, V_{IN} = 0 \leftrightarrow 5V$ -         0.40           Inductive Load (upper-lower arm)         -         0.30           Collector-emitter $V_{ar=}V_{are}$ -         -	Parameter         Condition         Min.         Typ.         Max.           Collector-emitter saturation voltage $V_D = V_{DB} = 15V, V_{IN} = 5V, \\ I_c = 75A$ $T_j = 25^{\circ}C$ -         1.50         1.90           FWDi forward voltage $V_{IN} = 0V, -I_C = 75A$ $T_j = 125^{\circ}C$ -         1.85         2.40           Switching times $V_{IN} = 0V, -I_C = 75A$ -         1.85         2.40 $V_{CC} = 600V, V_D = V_{DB} = 15V$ $I_c = 75A, T_j = 125^{\circ}C, V_{IN} = 0 \leftrightarrow 5V$ -         0.40         0.85 $I_c = 75A, T_j = 125^{\circ}C, V_{IN} = 0 \leftrightarrow 5V$ $I_c = 75A, T_j = 125^{\circ}C, V_{IN} = 0 \leftrightarrow 5V$ -         0.30         0.95           Inductive Load (upper-low = arm)         -         0.30         -         0.30         -           Collector-emitter $V_{or} = V_{ors}$ $V_{ors} = V_{ors}$ $I_c = 75A, T_c = 125^{\circ}C, T_c = 125$	

### CONVERTER PART

Symbol Parameter	Condition		Linit			
	Condition	Min.	Тур.	Max.	Unit	
I <sub>RRM</sub>	Repetitive reverse current	V <sub>R</sub> =V <sub>RRM</sub>	-	-	0.5	mA
VF	Forward voltage drop	I <sub>F</sub> =75A	-	1.30	1.75	V

#### **CONTROL (PROTECTION) PART**

Symbol	Parameter		Condition		Limits			
Symbol	Falameter		condition	Min.	Тур.	Max.	Unit	
			V <sub>D</sub> =15V, V <sub>IN</sub> =0V	-	-	4.70		
ID	Circuit current	Total of $V_{P1}$ - $V_{PC}$ , $V_{N1}$ - $V_{NC}$	V <sub>D</sub> =15V, V <sub>IN</sub> =5V	-	-	4.70	mA	
		Each part of V <sub>UFB</sub> -V <sub>UFS</sub> ,	V <sub>DB</sub> =15V, V <sub>IN</sub> =0V	-	-	2.40	IIIA	
I <sub>DB</sub>		VVFB-VVFS, VWFB-VWFS	V <sub>DB</sub> =15V, V <sub>IN</sub> =5V	-	-	2.40		
I <sub>SC</sub>	Short circuit trip level	Rs=30.1Ω (±1%), Not connecting outer shunt resistors to NU.NV.NW terminals (Note 5)		112.5	-	-	А	
UV <sub>DBt</sub>	P-side Control supply		Trip level	10.0	-	12.3	V	
$UV_{\text{DBr}}$	under-voltage protection(UV)		Reset level	10.4	-	12.9	V	
UV <sub>Dt</sub>	N-side Control supply		Trip level	10.3	-	12.5	V	
UV <sub>Dr</sub>	under-voltage protection(UV)		Reset level	10.8	-	13.0	V	
$V_{\text{FOH}}$		V <sub>SC</sub> = 0V, F <sub>o</sub> terminal pulle	ed up to 5V by 10kΩ	4.9	-	-	V	
V <sub>FOL</sub>	Fault output voltage	$V_{SC}$ = 1V, $I_{FO}$ = 1mA		-	-	0.95	V	
t <sub>FO</sub>	Fault output pulse width	C <sub>FO</sub> =22nF	(Note 6)	1.6	2.4	-	ms	
I <sub>IN</sub>	Input current	V <sub>IN</sub> = 5V		0.7	1.0	1.5	mA	
V <sub>th(on)</sub>	ON threshold voltage				-	2.6	V	
V <sub>th(off)</sub>	OFF threshold voltage	Applied between $U_P$ , $V_P$ , $W_P$ - $V_{NC}$ , $U_N$ , $V_N$ , $W_N$ - $V_{NC}$		0.8	-	-	V	
V <sub>OT</sub>	Temperature output	LVIC temperature=100°C, resistor=5.1kΩ	pull-down (Note 7)	2.89	3.02	3.14	V	

Note 5: Short circuit protection detects sense current divided from main current at N-side IGBT only. In the case that outer shunt resistor is inserted into main current path, protection current level Isc changes. For over -current protection by outside circuit, set the protection level under I<sub>CP</sub>.
6: Fault signal is output when short circuit or N-side control supply under-voltage protection works. The fault output pulse-width t<sub>FO</sub> depends on the capacitance of C<sub>FO</sub>. (C<sub>FO</sub> (typ.) = t<sub>FO</sub> x 9.1 x 10<sup>-6</sup>) [F])
7: DIPIPM doesn't shut down IGBTs and output fault signal automatically when temperature rises excessively. When temperature exceeds the protective level that

user defined, controller (MCU) should stop immediately. Temperature of LVIC vs.  $V_{0T}$  output characteristics is described in Fig.2. These minimum and maximum curves are based on theoretical designed value excluding LVIC temperature=100°C limits.

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## Fig. 2 Temperature of LVIC vs. Vot Output Characteristics

(These minimum and maximum curves are based on theoretical designed value excluding LVIC temperature=100°C limits.)

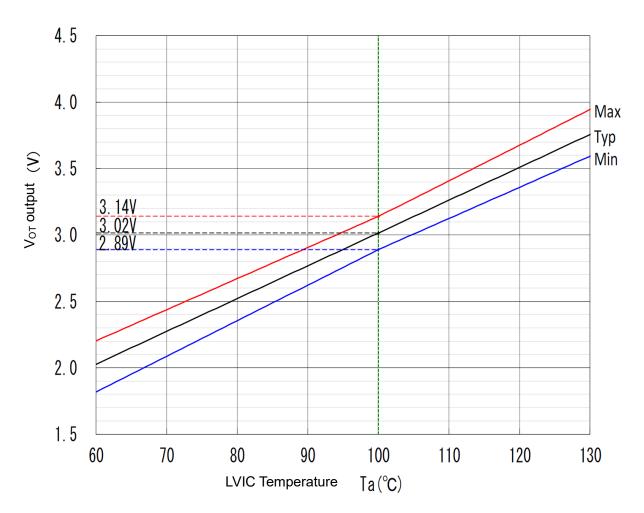
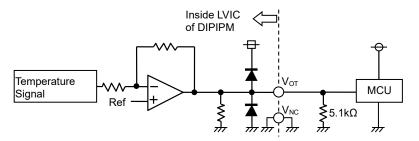


Fig. 3 Vot output circuit



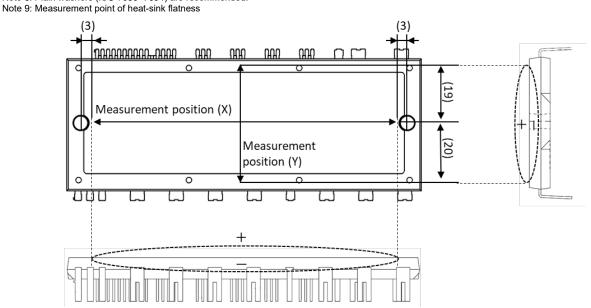
- (1) It is recommended to insert 5.1kΩ pull down resistor for getting linear output characteristics at low temperature (below room temperature). When the pull down resistor is inserted between V<sub>OT</sub> and V<sub>NC</sub> (control GND), the extra circuit current, which is calculated approximately by V<sub>OT</sub> output voltage divided by pull down resistance, flows as LVIC circuit current continuously. In the case of using V<sub>OT</sub> for detecting only higher temperature than room temperature, it isn't necessary to insert the pull down resistor.
- (2) In the case of using V<sub>OT</sub> with low voltage controller like 3.3V MCU, V<sub>OT</sub> output might exceed control supply voltage 3.3V when temperature rises excessively. If system uses low voltage controller, it is recommended to insert a clamp Di between control supply of the controller and V<sub>OT</sub> output for preventing over voltage destruction.
- (3) In the case of not using  $V_{\text{OT}},$  leave  $V_{\text{OT}}$  output NC (No Connection).

Refer the application note for this product about the usage of  $V_{\mbox{\scriptsize OT}}.$ 

# MECHANICAL CHARACTERISTICS AND RATINGS

Parameter	Condition		Reference	Limits			Unit
Falameter			Reference	Min.	Тур.	Max.	Unit
Mounting torque	Mounting screw : M4	(Note 8)	JEITA-ED-4701 402 method II	0.98	1.18	1.47	N∙m
Terminal strength pulling	Weight power terminal:40N control terminal:10N		JEITA-ED-4701 401 method I	10	-	-	s
Terminal strength bending	Load 10N, 90deg. bend		JEITA-ED-4701 401 method III	2	-	-	times
Weight				-	89	-	g
Heat radiation part flatness		(Note 9)		-50	-	130	μm

Note 8: Plain washers (ISO 7089~7094) are recommended.

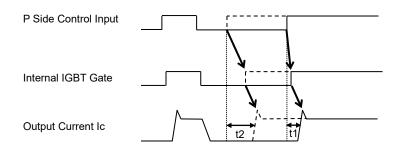


(Unit: mm)

#### **RECOMMENDED OPERATION CONDITIONS**

Sympol	Parameter	Condition		Unit		
Symbol	Parameter	Condition	Min.	Тур.	Max.	Unit
V <sub>cc</sub>	Supply voltage	Applied between P-NU, NV, NW	300	600	800	V
V <sub>D</sub>	Control supply voltage	Applied between V <sub>P1</sub> -V <sub>NC</sub> , V <sub>N1</sub> -V <sub>NC</sub>	13.5	15.0	16.5	V
V <sub>DB</sub>	Control supply voltage	Applied between V <sub>UFB</sub> -V <sub>UFS</sub> , V <sub>VFB</sub> -V <sub>VFS</sub> , V <sub>WFB</sub> -V <sub>WFS</sub>	13.0	15.0	18.5	V
$\Delta V_D, \Delta V_{DB}$	Control supply variation		-1	-	+1	V/µs
t <sub>dead</sub>	Arm shoot-through blocking time	For each input signal	2.5	-	-	μs
f <sub>PWM</sub>	PWM input frequency	T <sub>C</sub> ≤ 125°C, T <sub>j</sub> ≤ 150°C	-	-	20	kHz
V <sub>NC</sub>	V <sub>NC</sub> terminal voltage	Between V <sub>NC</sub> -NU, NV, NW (including surge)	-5.0	-	+5.0	V
PWIN(on)	Minimum input pulse width	(Note 10)	3.0	-	-	μs
PWIN(off)	winning in par pulse width	(Note 11)	3.0	-	-	μο

Note 10: DIPIPM might not make response if the input signal pulse width is less than PWIN(on). Note 11: DIPIPM might make no response or delayed response (P-side IGBT only) for input pulse width less than PWIN(off). Over rated collector current (Ic) operation, DIPIPM might make delayed response even if the input signal pulse width is PWIN(off) or more. The timing charts are described as below.

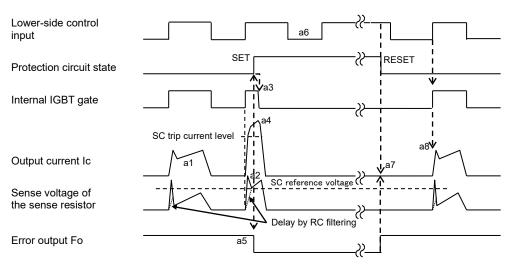


Real line: Normal response; turn on time t1 Broken line: Delayed response; turn on time t2 (t1: Normal switching time)

Fig. 4 Timing Charts of DIPIPM Protective Functions

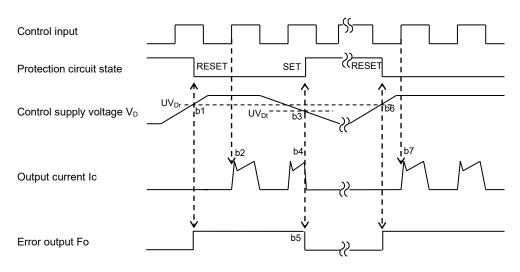
[A] Short-Circuit Protection (N-side only with the external sense resistor and RC filter)

- a1. Normal operation: IGBT ON and outputs current.
- a2. Short circuit current detection (SC trigger)
  - (It is recommended to set RC time constant 1.5~2.0µs so that IGBT shut down within 2.0µs when SC occurs.)
- a3. All N-side IGBT's gates are hard interrupted.
- a4. All N-side IGBTs turn OFF.
- a5.  $F_{\text{O}}$  outputs with a fixed pulse width determined by the external capacitor  $C_{\text{FO}}$
- a6. Input = "L": IGBT OFF
- a7. Fo finishes output, but IGBTs don't turn on until inputting next ON signal (L $\rightarrow$ H).
- (IGBT of each phase can return to normal state by inputting ON signal to each phase.) a8. Normal operation: IGBT ON and outputs current.



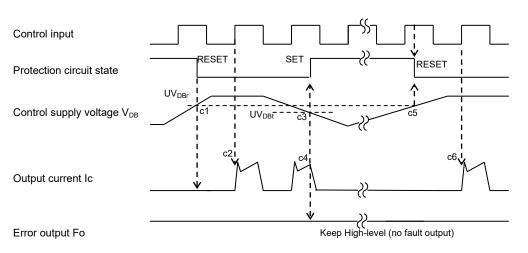
[B] Under-Voltage Protection (N-side, UV<sub>D</sub>)

- b1. Control supply voltage  $V_D$  exceeds under voltage reset level (UV<sub>Dr</sub>), but IGBT turns ON by next ON signal (L $\rightarrow$ H). (IGBT of each phase can return to normal state by inputting ON signal to each phase.)
- b2. Normal operation: IGBT ON and outputs current.
- b3.  $V_D$  level drops to under voltage trip level. (UV<sub>Dt</sub>).
- b4. All N-side IGBTs turn OFF in spite of control input condition.
- b5. Fo outputs for the period determined by the capacitance  $C_{FO}$ , but output is extended during  $V_D$  keeps below  $UV_{Dr}$ .
- b6.  $V_{\text{D}}$  level reaches  $UV_{\text{Dr}}.$
- b7. Normal operation: IGBT ON and outputs current by next ON signal (L $\rightarrow$ H).



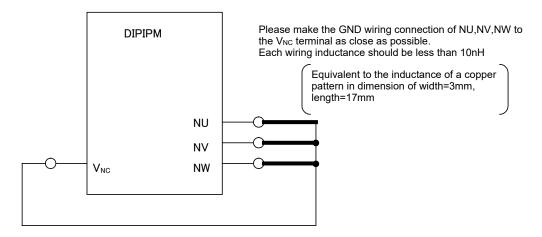
[C] Under-Voltage Protection (P-side, UVDB)

- c1. Control supply voltage  $V_{DB}$  rises. After the voltage reaches under voltage reset level  $UV_{DBr}$ , IGBT turns on by next ON signal (L $\rightarrow$ H).
- c2. Normal operation: IGBT ON and outputs current.
- c3.  $V_{DB}$  level drops to under voltage trip level (UV<sub>DBt</sub>).
- c4. IGBT of corresponding phase only turns OFF in spite of control input signal level, but there is no Fo signal output.
- c5.  $V_{DB}$  level reaches  $UV_{DBr}$ .
- c6. Normal operation: IGBT ON and outputs current by next ON signal (L $\rightarrow$ H).

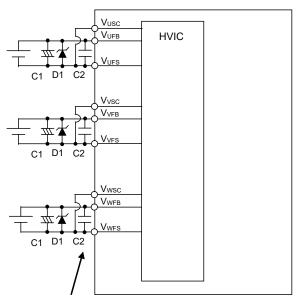


# **Peripheral circuits**

# 1) Pattern Wiring Around N terminal



# 2) Circuit Around Control supply voltage VDB

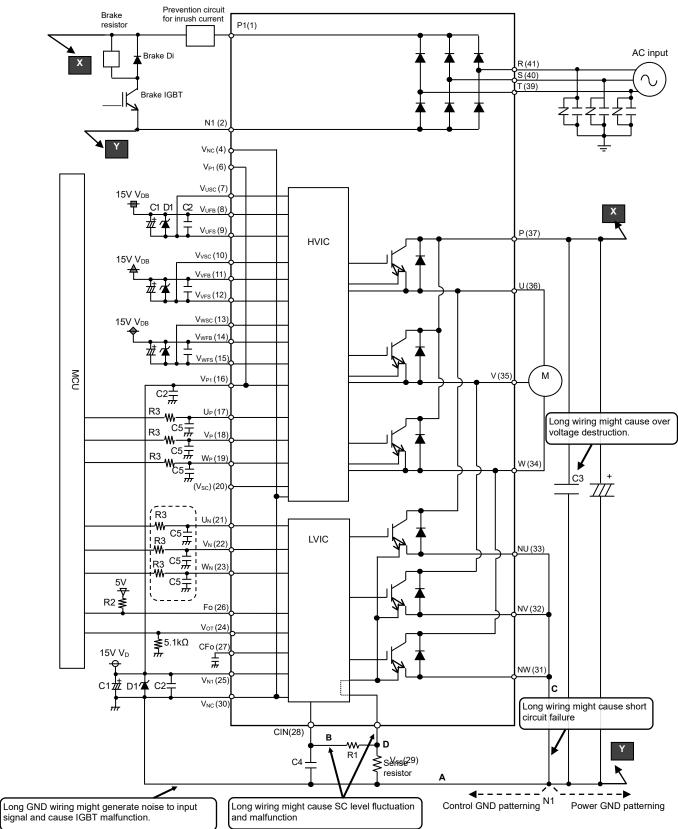


Capacitors C2 should be mounted as close to the terminals of the DIPIPM as possible.

- (1) All capacitors, especially C2 should be mounted as close to the terminals of the DIPIPM as possible. C1: good temperature, frequency characteristic electrolytic type, and C2:  $0.1\mu \sim 2\mu$ F, good temperature, frequency and DC bias characteristic type should be used (SMD type is recommended).
- (2) It is recommended to insert a Zener diode (24V,1W) between each pair of control supply terminals to prevent surge destruction.

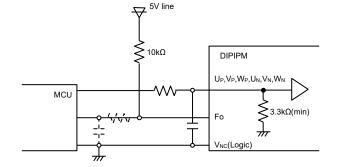
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Fig. 5 Example of Application Circuit



#### Note for the previous application circuit

- 1 :If control GND and power GND are patterned by common wiring, it may cause malfunction by fluctuation of power GND level. It is recommended to connect control GND and power GND at only a N1 point at which NU, NV, NW are connected to power GND line.
- 2 :It is recommended to insert a Zener diode D1 (24V/1W) between each pair of control supply terminals to prevent surge destruction.
- 3 :To prevent surge destruction, the wiring between the smoothing capacitor and the P, N1 terminals should be as short as possible. Also, insert a snubber capacitor C3 of appropriate capacity (1µF or more) between the P-N1 terminals. Please design the capacity of the snubber capacitor so that it is optimized according to the wiring pattern etc..
- 4 :R1, C4 of RC filter for preventing protection circuit malfunction is recommended to select tight tolerance, temp-compensated type. The time constant R1C4 should be set so that SC current is shut down within 2µs. (1.5µs~2µs is general value.) SC interrupting time might vary with the wiring pattern, so the enough evaluation on the real system is recommended. If R1 is too small, it may lead to delay of protection. So R1 should be min. 10 times larger resistance than Rs. (100 times is recommended.)
- 5 :To prevent erroneous operation, the wiring of A, B, C should be as short as possible.
- 6 :For sense resistor, the variation within 1% (including temperature characteristics), low inductance type is recommended. And the over 0.03W is recommended, but it is necessary to evaluate in your real system finally.
- 7 :To prevent erroneous SC protection, the wiring from V<sub>sc</sub> terminal to CIN filter should be divided at the point D that is close to the terminal of sense resistor. And the wiring should be patterned as short as possible.
- 8 :All capacitors should be mounted as close to the terminals of the DIPIPM as possible. (C1: good temperature, frequency characteristic electrolytic type, and C2: 0.1µ~2.0µF, good temperature, frequency and DC bias characteristic ceramic type are recommended.)
- 9 :Input drive is High-active type. There is a min. 3.3kΩ pull-down resistor in the input circuit of IC. To prevent malfunction, the wiring of each input should be as short as possible. And it is recommended to insert RC filter (e.g. R3=100Ω and C5=1000pF) and confirm the input signal level to meet the turn-on and turn-off threshold voltage. Thanks to HVIC inside the module, direct coupling to MCU without any opto-coupler or transformer isolation is possible.
- 10 :Fo output is open drain type. Fo output will be max  $0.95V(@I_{FO}=1mA,25^{\circ}C)$ , so it should be pulled up to MCU or control power supply (e.g. 5V,15V) by a resistor that makes I<sub>Fo</sub> up to 1mA. (In the case of pulled up to 5V, 10k $\Omega$  is recommended.)
- 11 :Error signal output width ( $t_{F_0}$ ) can be set by the capacitor connected to  $C_{F_0}$  terminal.  $C_{F_0}$  (typ.) =  $t_{F_0} \times 9.1 \times 10^{-6}$  (F)
- 12 :If high frequency noise superimposed to the control supply line, IC malfunction might happen and cause erroneous operation. To avoid such problem, voltage ripple of control supply line should meet dV/dt ≤+/-1V/μs, Vripple≤2Vp-p.
- 13 :For DIPIPM, it isn't recommended to drive same load by parallel connection with other phase IGBT or other DIPIPM.
- 14 :No.4 and No.30 V<sub>NC</sub> terminals (GND terminal for control supply) are connected mutually inside of DIPIPM+ and also No.6 and No.16 V<sub>P1</sub> terminals are connected mutually inside, please connect either No.4 or No.30 terminal to GND and also connect either No.6 or No.16 terminal to supply and make the unused terminal leave no connection.
- 15 :Please connect the  $V_{USC}-V_{UFS}$ (7-9 terminal),  $V_{VSC}-V_{VFS}$ (10-12 terminal),  $V_{WSC}-V_{WFS}$ (13-15terminal) externally.
- 16 :Although 3, 5, 20 and 38 terminals are dummy terminals, it may have an electric potential, so make it a no connection.



#### Fig. 6 MCU I/O Interface Circuit

Note)

Design for input RC filter depends on the PWM control scheme used in the application and the wiring impedance of the printed circuit board.

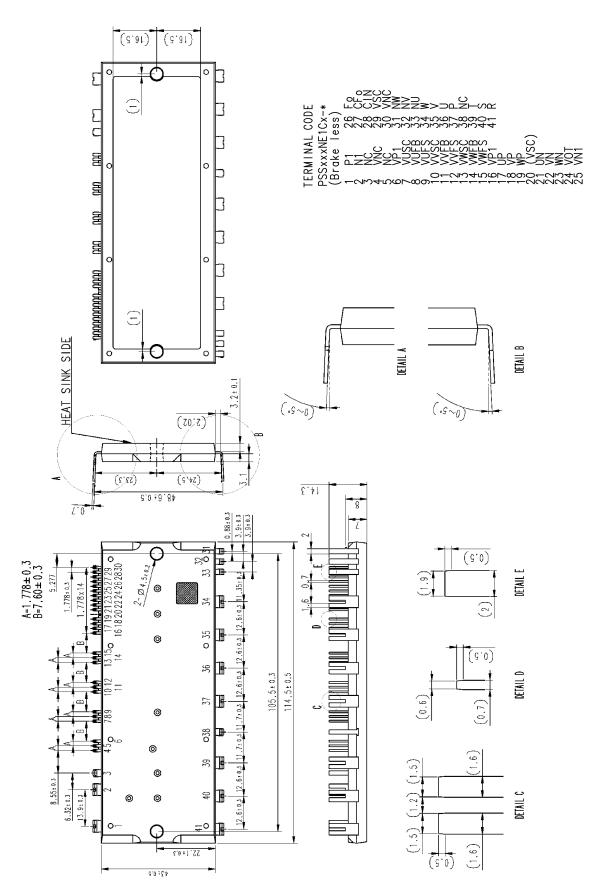
The DIPIPM input signal interface integrates a min.  $3.3k\Omega$  pull-down resistor. Therefore, when using RC filter, be careful to satisfy turn-on threshold voltage requirement.

Fo output is open drain type. It should be pulled up to the positive side of 5V or 15V power supply with the resistor that limits Fo sink current I<sub>Fo</sub> under 1mA. In the case of pulling up to 5V supply, over  $5.1k\Omega$  is needed. ( $10k\Omega$  is recommended.)

# < DIPIPM > **PSS75NE1CT** TRANSFER MOLDING TYPE INSULATED TYPE

Fig. 7 Package Outlines

Dimensions in mm



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